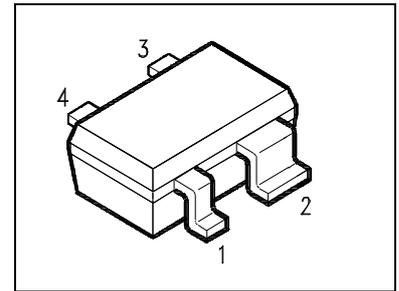


## NPN Silicon RF Transistor

- For High Gain Low Noise Amplifiers
- For Oscillators up to 10 GHz
- Noise Figure  $F = 1.05$  dB at 1.8 GHz  
Outstanding  $G_{ms} = 20$  dB at 1.8 GHz
- Transition Frequency  $f_T = 25$  GHz
- Gold metalization for high reliability
- **SIEGET<sup>®</sup>25-Line**  
**Siemens G**rounded **E**mitter Transistor-  
**25** GHz  $f_T$ -**L**ine



**ESD:** Electrostatic discharge sensitive device, observe handling precautions!

Type	Marking	Ordering Code (8-mm taped)	Pin Configuration				Package <sup>1)</sup>
			1	2	3	4	
BFP420	AMs	Q62702-F1591	B	E	C	E	SOT343

### Maximum Ratings

Parameter	Symbol		Unit
Collector-emitter voltage	$V_{CEO}$	4.5	V
Collector-base voltage	$V_{CBO}$	15	V
Emitter-base voltage	$V_{EBO}$	1.5	V
Collector current	$I_C$	35	mA
Base current	$I_B$	3	mA
Total power dissipation, $T_s \leq 107^\circ\text{C}$ <sup>2)3)</sup>	$P_{tot}$	160	mW
Junction temperature	$T_j$	150	$^\circ\text{C}$
Ambient temperature range	$T_A$	-65...+150 $^\circ\text{C}$	$^\circ\text{C}$
Storage temperature range	$T_{stg}$	-65...+150 $^\circ\text{C}$	$^\circ\text{C}$

### Thermal Resistance

Junction-soldering point <sup>2)</sup>	$R_{th JS}$	270	K/W
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1) For detailed information see chapter Package

2)  $T_S$  is measured on the emitter lead at the soldering point to the pcb.

3)  $P_{tot}$  due to Maximum Ratings.

At typical  $T_s \leq 80^\circ\text{C}$ :  $P_{tot} = 250$  mW due to thermal characteristics.

**Electrical Characteristics**

at  $T_A = 25\text{ °C}$ , unless otherwise specified.

Parameter	Symbol	Value			Unit
		min.	typ.	max.	

**DC Characteristics**

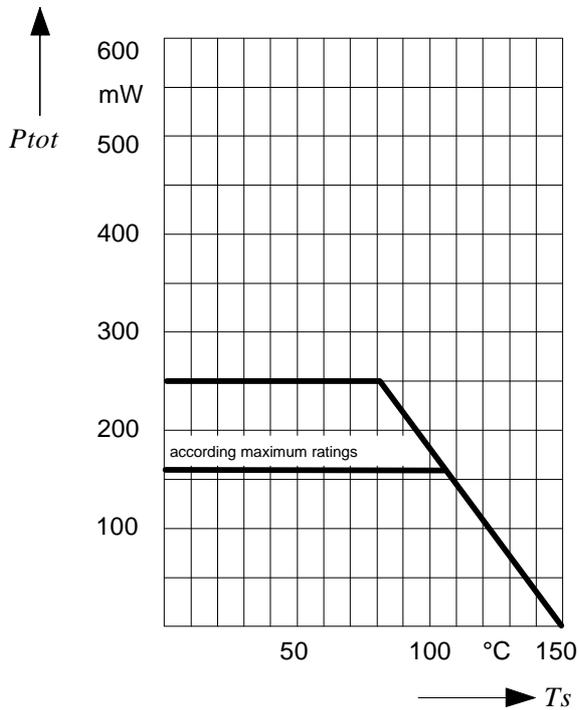
Collector-emitter breakdown voltage $I_C = 1\text{ mA}$	$V_{(BR)CEO}$	4.5	5	6.5	V
Collector-cutoff current $V_{CB} = 5\text{ V}, I_E = 0$	$I_{CBO}$	-	-	200	nA
Emitter base cutoff current $V_{EB} = 1.5\text{ V}, I_C = 0$	$I_{EBO}$	-	-	35	$\mu\text{A}$
DC current gain $I_C = 20\text{ mA}, V_{CE} = 4\text{ V}$	$h_{FE}$	50	80	150	

**AC Characteristics**

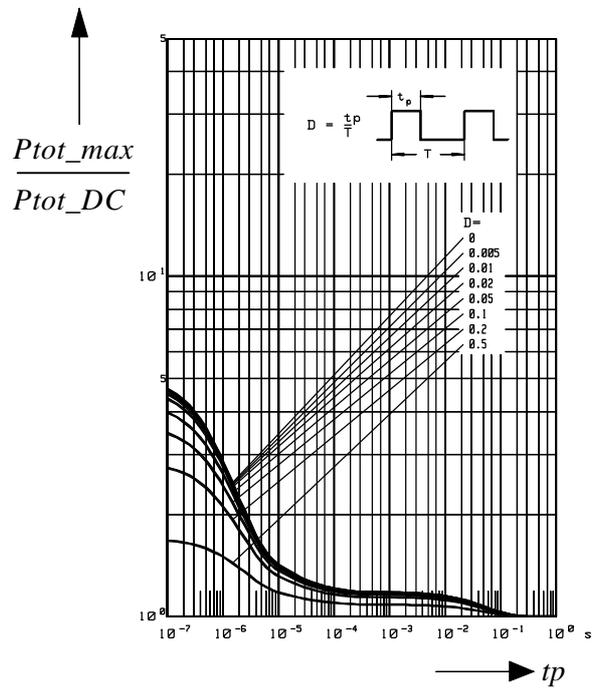
Transition frequency $I_C = 30\text{ mA}, V_{CE} = 3\text{ V}, f = 2\text{ GHz}$	$f_T$	20	25	-	GHz
Collector-base capacitance $V_{CB} = 2\text{ V}, V_{BE} = v_{be} = 0, f = 1\text{ MHz}$	$C_{cb}$	-	0.15	0.24	pF
Collector-emitter capacitance $V_{CE} = 2\text{ V}, V_{BE} = v_{be} = 0, f = 1\text{ MHz}$	$C_{ce}$	-	0.41	-	pF
Emitter-base capacitance $V_{EB} = 0.5\text{ V}, V_{CB} = v_{cb} = 0, f = 1\text{ MHz}$	$C_{eb}$	-	0.55	-	pF
Noise figure $I_C = 5\text{ mA}, V_{CE} = 2\text{ V}, f = 1.8\text{ GHz}, Z_S = Z_{Sopt}$	$F$	-	1.05	1.4	dB
Power gain $I_C = 20\text{ mA}, V_{CE} = 2\text{ V}, f = 1.8\text{ GHz}, Z_S = Z_{Sopt}, Z_L = Z_{Lopt}$	$G_{ms}^{1)}$	-	20	-	dB
Insertion power gain $I_C = 20\text{ mA}, V_{CE} = 2\text{ V}, f = 1.8\text{ GHz}, Z_S = Z_L = 50\Omega$	$ S_{21} ^2$	14	17	-	dB
Third order intercept point at output $I_C = 20\text{ mA}, V_{CE} = 2\text{ V}, f = 1.8\text{ GHz}, Z_S = Z_{Sopt}, Z_L = Z_{Lopt}$	$IP_3$	-	22	-	dBm
1dB Compression point $I_C = 20\text{ mA}, V_{CE} = 2\text{ V}, f = 1.8\text{ GHz}, Z_S = Z_{Sopt}, Z_L = Z_{Lopt}$	$P_{-1dB}$	-	12	-	dBm

1)  $G_{ms} = \left| \frac{S_{21}}{S_{12}} \right|$

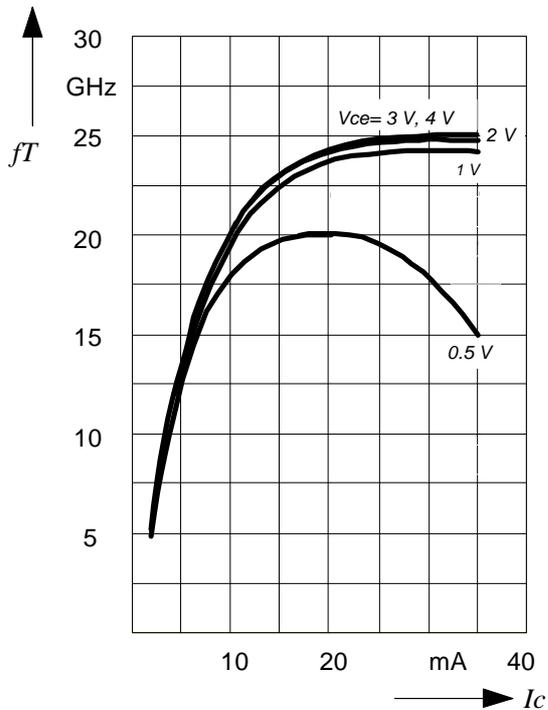
**Total Power Dissipation**  
versus Soldering Point Temperature



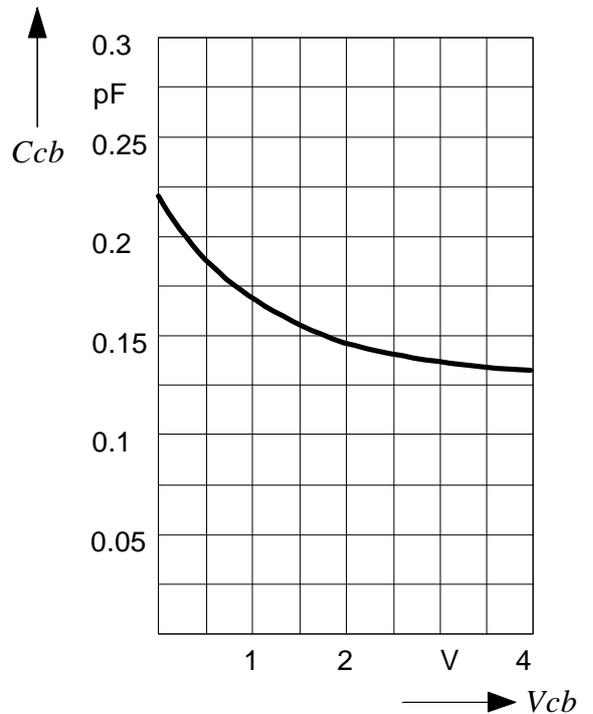
**Permissible Pulse Power Dissipation**  
versus On-Time ( $V_{CE0max} = 4.5 V$ )



**Transition Frequency**  
versus Collector Current  
 $f = 2 GHz$

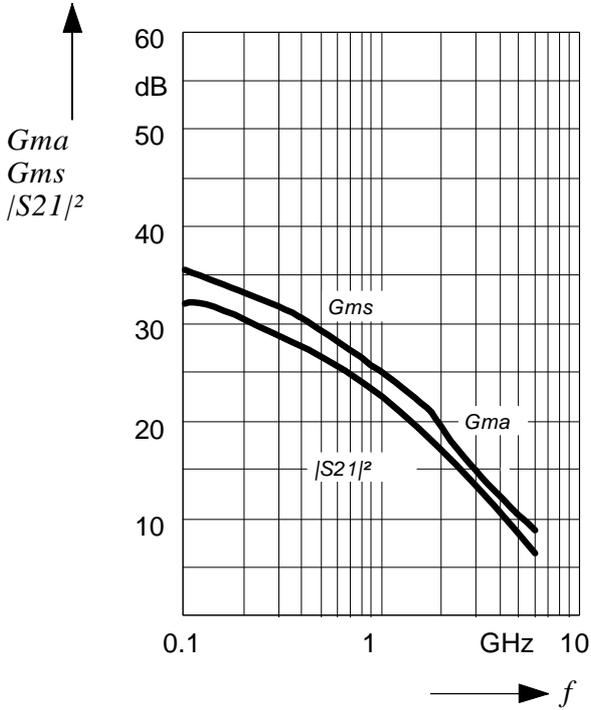


**Collector-base Capacitance**  
versus Collector-base Voltage  
 $V_{BE} = 0 V, f = 1 MHz$



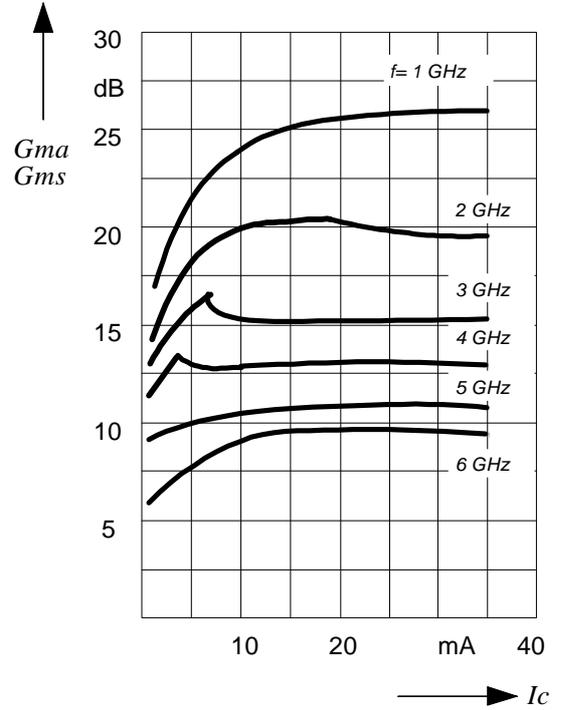
**Power Gain**

versus Frequency  
 $V_{CE} = 2 \text{ V}$ ,  $I_C = 20 \text{ mA}$



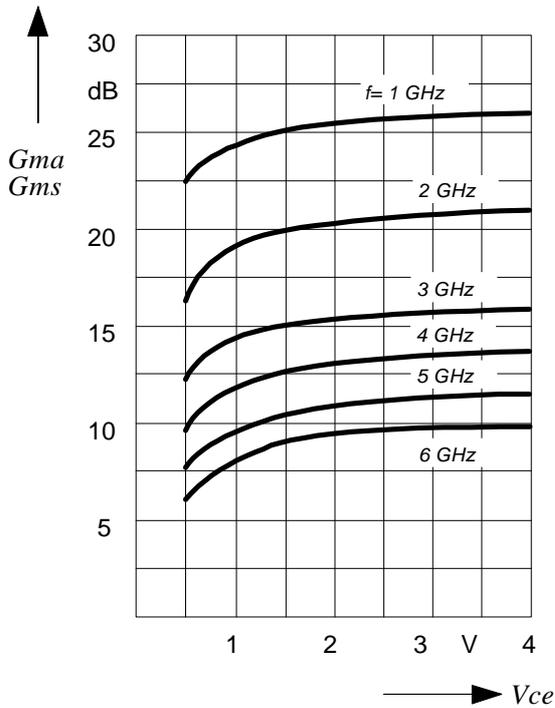
**Power Gain**

versus Collector Current  
 $V_{CE} = 2 \text{ V}$



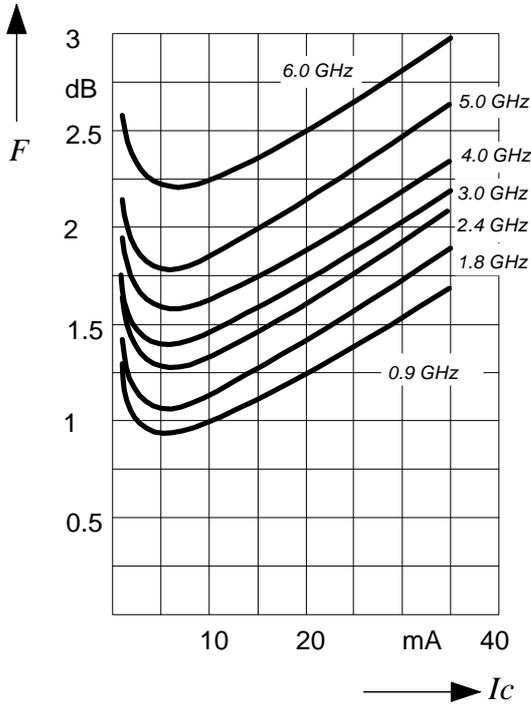
**Power Gain**

versus Collector Voltage  
 $I_C = 20 \text{ mA}$



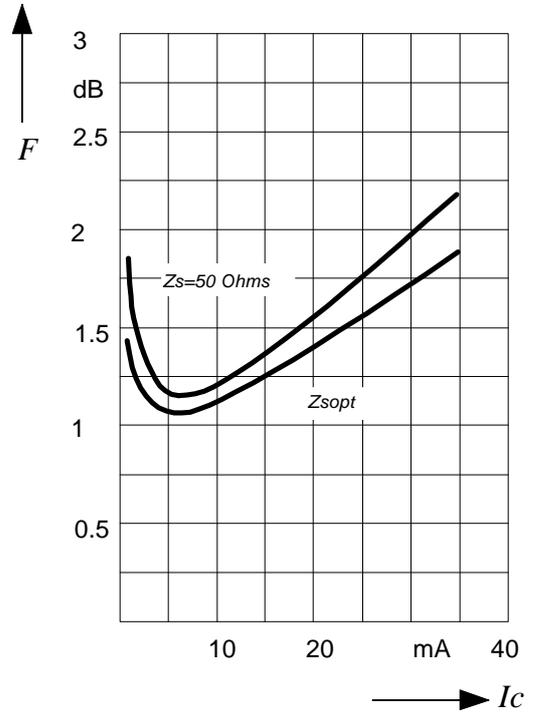
**Noise Figure**

versus Collector Current  
 $V_{CE} = 2\text{ V}$ ,  $Z_S = Z_{Sopt}$



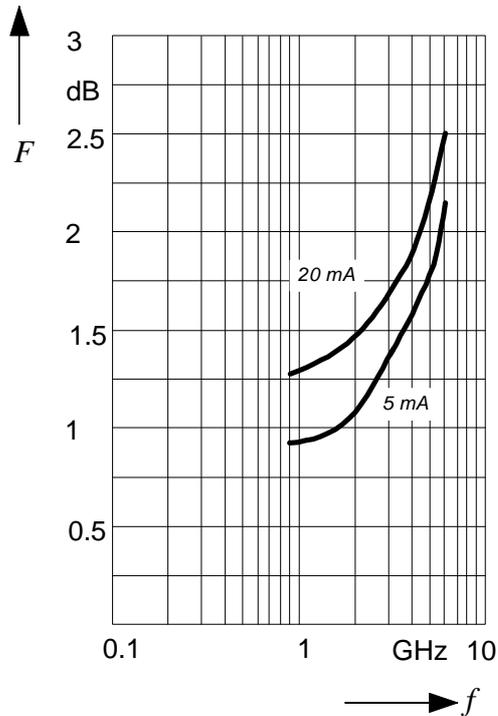
**Noise Figure**

versus Collector Current  
 $V_{CE} = 2\text{ V}$ ,  $f = 1.8\text{ GHz}$



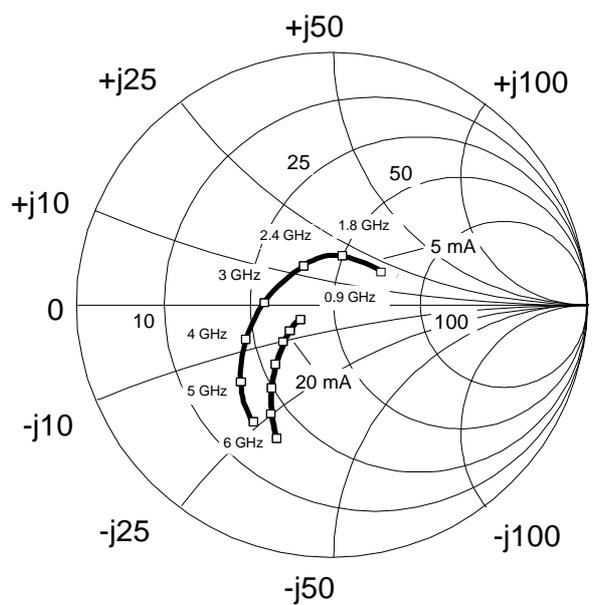
**Noise Figure versus Frequency**

$V_{CE} = 2\text{ V}$ ,  $I_C = 5\text{ mA} / 20\text{ mA}$ ,  
 $Z_S = Z_{Sopt}$



**Source Impedance for min.**

Noise Figure versus Frequency  
 $V_{CE} = 2\text{ V}$ ,  $I_C = 5\text{ mA} / 20\text{ mA}$



**Common Emitter S-Parameters**

$f$	$S_{11}$		$S_{21}$		$S_{12}$		$S_{22}$	
GHz	MAG	ANG	MAG	ANG	MAG	ANG	MAG	ANG
$V_{CE} = 2\text{ V}, I_C = 20\text{ mA}$								
0.01	0.452	-2.3	37.62	178.3	0.0011	94.4	0.956	-0.6
0.1	0.447	-25.1	36.30	164.7	0.0068	82.5	0.941	-12.4
0.5	0.386	-101.1	23.41	121.0	0.0262	61.7	0.632	-47.2
1.0	0.378	-146.2	13.99	96.0	0.0395	57.8	0.395	-63.9
2.0	0.405	173.5	7.18	70.8	0.0664	54.0	0.222	-87.3
3.0	0.446	149.4	4.77	52.6	0.0949	47.1	0.133	-111.3
4.0	0.501	130.0	3.52	36.8	0.1206	38.5	0.133	-158.5
6.0	0.599	104.8	2.27	8.2	0.1646	18.9	0.196	142.0
8.0	0.700	78.5	1.51	-20.8	0.1800	-2.4	0.289	99.3
9.0	0.758	67.6	1.25	-34.4	0.1820	-13.0	0.379	84.1
10.0	0.800	62.0	1.04	-43.5	0.1800	-19.3	0.465	76.6

$V_{CE} = 2\text{ V}, I_C = 5\text{ mA}$

0.01	0.790	-1.0	15.14	179.2	0.0012	83.4	0.988	-0.7
0.1	0.786	-11.6	14.98	171.8	0.0092	84.1	0.982	-6.5
0.5	0.702	-55.7	12.86	140.1	0.0398	62.8	0.857	-29.8
1.0	0.589	-99.1	9.63	112.6	0.0603	46.5	0.647	-48.6
2.0	0.507	-156.0	5.60	79.4	0.0798	34.6	0.401	-70.3
3.0	0.511	168.5	3.84	57.1	0.0957	29.8	0.267	-84.2
4.0	0.549	142.0	2.87	38.5	0.1121	25.1	0.207	-110.5
5.0	0.604	123.9	2.26	22.1	0.1285	19.4	0.150	-137.3
6.0	0.633	110.0	1.86	6.7	0.1442	13.1	0.173	-169.8

**Common Emitter Noise Parameters**

$f$	$F_{min}^{1)}$	$G_a^{1)}$	$\Gamma_{opt}$		$R_N$	$r_n$	$F_{50\Omega}^{2)}$	$ S_{21} ^2^{2)}$
GHz	dB	dB	MAG	ANG	$\Omega$	-	dB	dB
$V_{CE} = 2\text{ V}, I_C = 5\text{ mA}$								
0.9	0.90	20.5	0.28	41.0	8.7	0.17	1.02	20.3
1.8	1.05	15.2	0.20	82.0	6.7	0.13	1.11	15.8
2.4	1.25	13.0	0.20	124.0	5.5	0.11	1.32	13.5
3.0	1.38	12.1	0.22	-175.0	5.0	0.10	1.48	11.6
4.0	1.55	10.3	0.33	-157.0	5.5	0.11	1.83	9.1
5.0	1.75	8.6	0.45	-142.0	5.0	0.10	2.20	7.0
6.0	2.20	6.4	0.53	-123.0	15.0	0.30	3.30	5.3

1) Input matched for minimum noise figure, output for maximum gain

2)  $Z_S = Z_L = 50\Omega$

For more and detailed S- and Noise-parameters please contact your local Siemens distributor or sales office to obtain a SIEMENS Application Notes CD-ROM or see Internet:

<http://www.siemens.de/Semiconductor/products/35/357.htm>

### SPICE Parameters:

#### Transistor Chip Data T502 (Berkeley-SPICE 2G.6 Syntax):

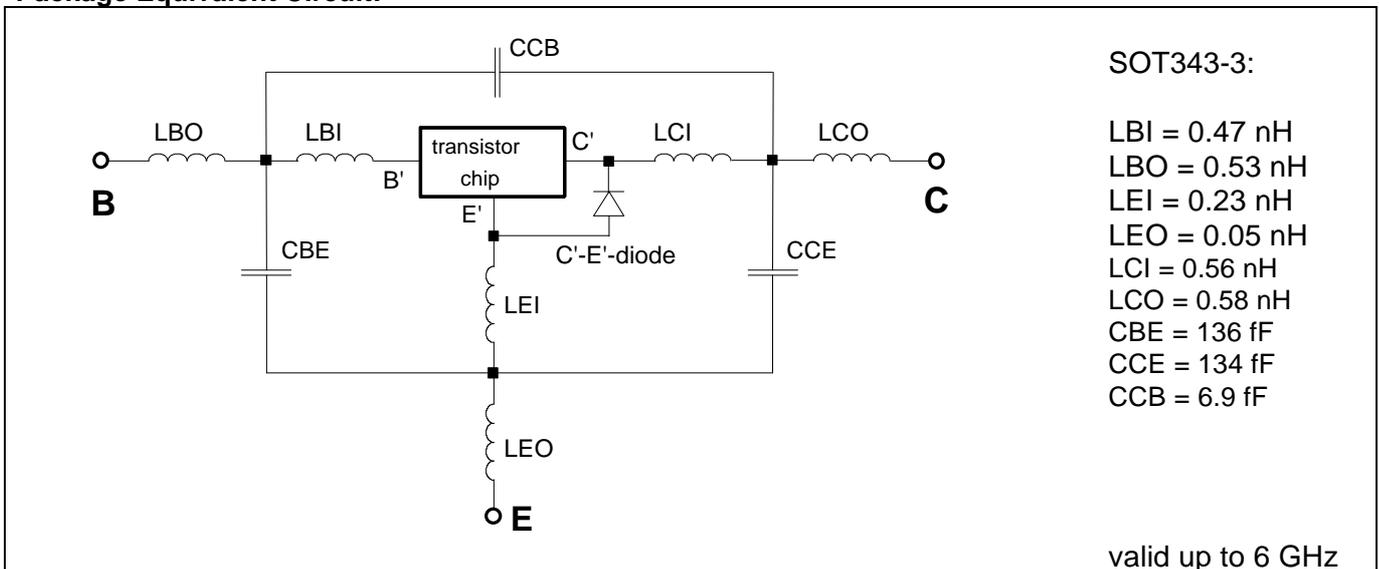
IS =	0.20045	fA	BF =	72.534	-	NF =	1.2432	-
VAF =	28.383	V	IKF =	0.48731	A	ISE =	19.049	fA
NE =	2.0518	-	BR =	7.8287	-	NR =	1.3325	-
VAR =	19.705	V	IKR =	0.69141	A	ISC =	0.019237	fA
NC =	1.1724	-	RB =	3.4849	OHM	IRB =	0.72983	mA
RBM =	8.5757	OHM	RE =	0.31111	OHM	RC =	0.10105	OHM
CJE =	1.8063	fF	VJE =	0.8051	V	MJE =	0.46576	-
TF =	6.7661	ps	XTF =	0.42199	-	VTF =	0.23794	V
ITF =	1.0	mA	PTF =	0	deg	CJC =	234.53	fF
VJC =	0.81969	V	MJC =	0.30232	-	XCJC =	0.3	-
TR =	2.3249	ns	CJS =	0	fF	VJS =	0.75	V
MJS =	0	-	XTB =	0	-	EG =	1.11	eV
XTI =	3.0	-	FC =	0.73234	-	Tnom =	300	K

#### C'-E'- Diode Data (Berkeley-SPICE 2G.6 Syntax):

IS =	3.5	fA	N =	1.02	-	RS =	10	$\Omega$
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All parameters are ready to use, no scaling is necessary.

#### Package Equivalent Circuit:



The SOT343 package has two emitter leads. To avoid high complexity of the package equivalent circuit, both leads are combined in one electrical connection.

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 Institut für Mobil- und Satellitenfunktechnik (IMST)  
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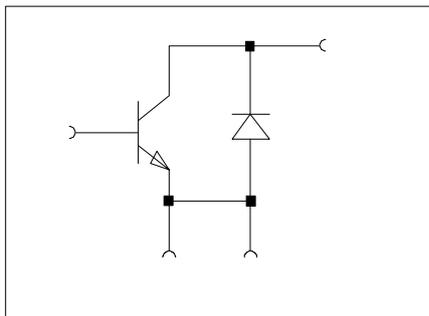
For more examples and ready to use parameters please contact your local Siemens distributor or sales office to obtain a SIEMENS Application Notes CD-ROM or see Internet:  
<http://www.siemens.de/Semiconductor/products/35/357.htm>

For non-linear simulation:

- Use transistor chip parameters in Berkeley SPICE 2G.6 syntax for all simulators.
- If you need simulation of the reverse characteristics, add the diode with the C'-E'-diode data between collector and emitter.
- Simulation of the package is not necessary for frequencies < 100 MHz.  
For higher frequencies add the wiring of the package equivalent circuit around the non-linear transistor and diode model.

Note:

- This transistor is constructed in a common emitter configuration. This feature causes an additional, reverse biased diode between emitter and collector, which does not effect normal operation.



**Transistor Schematic Diagram**

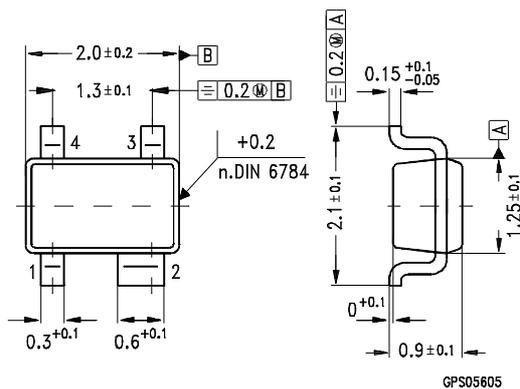
The common emitter configuration shows the following advantages:

- Higher gain because of lower emitter inductance.
- Power is dissipated via the grounded emitter leads, because the chip is mounted on the copper emitter leadframe.

Please note, that the broadest lead is the emitter lead.

- The AC-Characteristics are verified by random sampling.

**Package**



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